### 5.0 OR 3.0V, 512 bit ( $64 \times 8$ ) SERIAL RTC and NVRAM SUPERVISOR

FEATURES SUMMARY
■ 5.0 OR 3.0V OPERATING VOLTAGE

- SERIAL INTERFACE SUPPORTS ${ }^{2} \mathrm{C}$ BUS ( 400 KHz )
- NVRAM SUPERVISOR FOR EXTERNAL LPSRAM
- OPTIMIZED FOR MINIMAL INTERCONNECT TO MCU
■ 2.5 TO 5.5V OSCILLATOR OPERATING VOLTAGE

■ AUTOMATIC SWITCH-OVER and DESELECT CIRCUITRY

■ CHOICE OF POWER-FAIL DESELECT VOLTAGES

- M41ST85Y: $\mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V ;
$4.20 \mathrm{~V} \leq \mathrm{V}_{\mathrm{PFD}} \leq 4.50 \mathrm{~V}$
- M41ST85W: $\mathrm{V}_{\mathrm{CC}}=2.7$ to 3.6 V ;
$2.55 \mathrm{~V} \leq \mathrm{V}_{\text {PFD }} \leq 2.70 \mathrm{~V}$
- 1.25V REFERENCE (for PFI/PFO)

■ COUNTERS FOR TENTHS/HUNDREDTHS OF SECONDS, SECONDS, MINUTES, HOURS, DAY, DATE, MONTH, YEAR, and CENTURY

- 44 BYTES OF GENERAL PURPOSE RAM
- PROGRAMMABLE ALARM and INTERRUPT FUNCTION (VALID EVEN DURING BATTERY BACK-UP MODE)
- WATCHDOG TIMER
- MICROPROCESSOR POWER-ON RESET
- BATTERY LOW FLAG
- POWER-DOWN TIMESTAMP (HT BIT)

■ ULTRA-LOW BATTERY SUPPLY CURRENT OF 500nA (MAX)

- PACKAGING INCLUDES A 28-LEAD SOIC and SNAPHAT ${ }^{\circledR}$ TOP (to be Ordered Separately)
■ SOIC SNAPHAT PACKAGE PROVIDES DIRECT CONNECTION FOR A SNAPHAT TOP WHICH CONTAINS THE BATTERY and CRYSTAL

■ SOIC EMBEDDED CRYSTAL PACKAGE (MX) OPTION

Figure 1. 28-pin SOIC Package


Figure 2. 28-pin ( 300 mil ) SOIC Package
EMBEDDED Crystal

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## SUMMARY DESCRIPTION

The M41ST85Y/W Serial TIMEKEEPER ${ }^{\circledR} /$ /Controller SRAM is a low power 512-bit, static CMOS SRAM organized as 64 words by 8 bits. A built-in 32.768 kHz oscillator (external crystal controlled) and 8 bytes of the SRAM (see Table 8, page 18) are used for the clock/calendar function and are configured in binary coded decimal (BCD) format.
An additional 12 bytes of RAM provide status/control of Alarm, Watchdog and Square Wave functions. Addresses and data are transferred serially via a two line, bi-directional $1^{2} \mathrm{C}$ interface. The built-in address register is incremented automatically after each WRITE or READ data byte. The M41ST85Y/W has a built-in power sense circuit which detects power failures and automatically switches to the battery supply when a power failure occurs. The energy needed to sustain the SRAM and clock operations can be supplied by a small lithium button-cell supply when a power failure occurs.
Functions available to the user include a non-volatile, time-of-day clock/calendar, Alarm interrupts, Watchdog Timer and programmable Square Wave output. Other features include a Power-On Reset as well as two additional debounced inputs ( $\overline{\text { RSTIN1 }}$ and RSTIN2) which can also generate an output Reset (RST). The eight clock address locations contain the century, year, month, date, day, hour, minute, second and tenths/hundredths of a second in 24 hour BCD format. Corrections for 28,

29 (leap year - valid until year 2100), 30 and 31 day months are made automatically.
The M41ST85Y/W is supplied in a 28 -lead SOIC SNAPHAT ${ }^{\circledR}$ package (which integrates both crystal and battery in a single SNAPHAT top) or a 28pin, 300mil SOIC package (MX) which includes an embedded 32 kHz crystal.
The 28 -pin, 330 mil SOIC provides sockets with gold plated contacts at both ends for direct connection to a separate SNAPHAT housing containing the battery and crystal. The unique design allows the SNAPHAT battery/crystal package to be mounted on top of the SOIC package after the completion of the surface mount process.
Insertion of the SNAPHAT housing after reflow prevents potential battery and crystal damage due to the high temperatures required for device sur-face-mounting. The SNAPHAT housing is also keyed to prevent reverse insertion.
The SOIC and battery/crystal packages are shipped separately in plastic anti-static tubes or in Tape \& Reel form. For the 28-lead SOIC, the battery/crystal package (e.g., SNAPHAT) part number is "M4TXX-BR12SH" (see Table 15, page 27).
Caution: Do not place the SNAPHAT battery/crystal top in conductive foam, as this will drain the lithium button-cell battery.
The 300 mil, embedded crystal SOIC requires only a user-supplied battery to provide non-volatile operation.

Figure 3. Logic Diagram


Note: 1. For 28 -pin, 300 mil embedded crystal SOIC only.

Figure 4. 28-pin SOIC Connections


Table 1. Signal Names

| $\overline{\mathrm{E}}_{\text {CON }}$ | Conditioned Chip Enable Output |
| :--- | :--- |
| $\overline{\mathrm{EX}}$ | External Chip Enable |
| $\overline{\mathrm{RQ} / \text { FT/OUT }}$ | Interrupt/Frequency Test/Out Output <br> (Open Drain) |
| PFI | Power Fail Input |
| $\overline{\text { PFO }}$ | Power Fail Output |
| $\overline{\text { RST }}$ | Reset Output (Open Drain) |
| $\overline{\text { RSTIN1 }}$ | Reset 1 Input |
| $\overline{\text { RSTIN2 }}$ | Reset 2 Input |
| SCL | Serial Clock Input |
| SDA | Serial Data Input/Output |
| SQW | Square Wave Output |
| WDI | Watchdog Input |
| VCC | Supply Voltage |
| V $_{\text {OUT }}$ | Voltage Output |
| VSS | Ground |
| V $_{\text {BAT }}{ }^{(1)}$ | Battery Supply Voltage |

Note: 1. For 28-pin, 300mil embedded crystal SOIC only.

Figure 5. 28-pin, 300mil SOIC (MX) Connections


Figure 6. Block Diagram


Note: 1. Open drain output
2. Integrated into SOIC package for MX package option.

Figure 7. Hardware Hookup


Note: 1. Required for embedded crystal (MX) package only.

## MAXIMUM RATING

Stressing the device above the rating listed in the "Absolute Maximum Ratings" table may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is
not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality documents.

Table 2. Absolute Maximum Ratings

| Symbol | Parameter |  | Value | Unit |
| :---: | :---: | :---: | :---: | :---: |
| TSTG | Storage Temperature (Vcc Off, Oscillator Off) | SNAPHAT ${ }^{\text {® }}$ | -40 to 85 | ${ }^{\circ} \mathrm{C}$ |
|  |  | SOIC | -55 to 125 | ${ }^{\circ} \mathrm{C}$ |
| TSLD ${ }^{(1)}$ | Lead Solder Temperature for 10 seconds |  | 260 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{10}$ | Input or Output Voltage |  | -0.3 to $\mathrm{V}_{\mathrm{CC}}+0.3$ | v |
| VCC | Supply Voltage | M41ST85Y | -0.3 to 7 | V |
|  |  | M41ST85W | -0.3 to 4.6 | V |
| 10 | Output Current |  | 20 | mA |
| $\mathrm{P}_{\mathrm{D}}$ | Power Dissipation |  | 1 | W |

Note: 1. Reflow at peak temperature of $215^{\circ} \mathrm{C}$ to $225^{\circ} \mathrm{C}$ for $<60$ seconds (total thermal budget not to exceed $180^{\circ} \mathrm{C}$ for between 90 to 120 seconds).

CAUTION: Negative undershoots below -0.3 V are not allowed on any pin while in the Battery Back-up mode.
CAUTION: Do NOT wave solder SOIC to avoid damaging SNAPHAT sockets.

## DC AND AC PARAMETERS

This section summarizes the operating and measurement conditions, as well as the DC and AC characteristics of the device. The parameters in the following DC and AC Characteristic tables are derived from tests performed under the Measure-
ment Conditions listed in the relevant tables. Designers should check that the operating conditions in their projects match the measurement conditions when using the quoted parameters.

Table 3. DC and AC Measurement Conditions

| Parameter | M41ST85Y | M41ST85W |
| :--- | :---: | :---: |
| V ${ }_{\text {CC }}$ Supply Voltage | 4.5 to 5.5 V | 2.7 to 3.6 V |
| Ambient Operating Temperature | -40 to $85^{\circ} \mathrm{C}$ | -40 to $85^{\circ} \mathrm{C}$ |
| Load Capacitance (CL) | 100 pF | 50 pF |
| Input Rise and Fall Times | $\leq 50 \mathrm{~ns}$ | $\leq 50 \mathrm{~ns}$ |
| Input Pulse Voltages | 0.2 to $0.8 \mathrm{~V}_{\mathrm{CC}}$ | 0.2 to $0.8 \mathrm{~V}_{\mathrm{CC}}$ |
| Input and Output Timing Ref. Voltages | 0.3 to $0.7 \mathrm{~V}_{\mathrm{CC}}$ | 0.3 to $0.7 \mathrm{~V}_{\mathrm{CC}}$ |

Note: Output High Z is defined as the point where data is no longer driven.
Figure 8. AC Testing Input/Output Waveforms


Note: 50 pF for M41ST85W.
Table 4. Capacitance

| Symbol | Parameter $^{(1,2)}$ | Min | Max | Unit |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{C}_{\text {IN }}$ | Input Capacitance |  | 7 | pF |
| Cout $^{(3)}$ | Output Capacitance |  | 10 | pF |
| $\mathrm{t}_{\mathrm{LP}}$ | Low-pass filter input time constant (SDA and SCL) |  | 50 | ns |

Note: 1. Effective capacitance measured with power supply at 5V. Sampled only, not $100 \%$ tested.
2. At $25^{\circ} \mathrm{C}, \mathrm{f}=1 \mathrm{MHz}$.
3. Outputs are deselected.

Table 5. DC Characteristics

| Sym | Parameter | Test Condition ${ }^{(1)}$ | M41ST85Y |  |  | M41ST85W |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min | Typ | Max | Min | Typ | Max |  |
| $\mathrm{I}_{\mathrm{BAT}}{ }^{(2)}$ | Battery Current OSC ON | $\begin{aligned} & \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \\ & \mathrm{~V}_{\mathrm{CC}}=0 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{BAT}}=3 \mathrm{~V} \end{aligned}$ |  | 400 | 500 |  | 400 | 500 | nA |
|  | Battery Current OSC OFF |  |  | 50 |  |  | 50 |  | nA |
| ICC1 | Supply Current | $\mathrm{f}=400 \mathrm{kHz}$ |  |  | 1.4 |  |  | 0.75 | mA |
| ICC2 | Supply Current (Standby) | $\begin{aligned} & \mathrm{SCL}, \mathrm{SDA}= \\ & \mathrm{V}_{\mathrm{CC}}-0.3 \mathrm{~V} \end{aligned}$ |  |  | 1 |  |  | 0.50 | mA |
| $\mathrm{ILI}^{(3)}$ | Input Leakage Current | $\begin{gathered} 0 \mathrm{~V} \leq \mathrm{V}_{\text {IN }} \leq \\ \mathrm{V}_{\mathrm{CC}} \end{gathered}$ |  |  | $\pm 1$ |  |  | $\pm 1$ | $\mu \mathrm{A}$ |
|  | Input Leakage Current (PFI) |  | -25 | 2 | 25 | -25 | 2 | 25 | nA |
| $\mathrm{ILO}^{(4)}$ | Output Leakage Current | $\begin{gathered} 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{IN}} \leq \\ \mathrm{V}_{\mathrm{CC}} \end{gathered}$ |  |  | $\pm 1$ |  |  | $\pm 1$ | $\mu \mathrm{A}$ |
| lout1 ${ }^{(5)}$ | Vout Current (Active) | $\begin{gathered} \mathrm{V}_{\text {OUT1 }}> \\ \mathrm{V}_{\mathrm{CC}}-0.3 \mathrm{~V} \end{gathered}$ |  |  | 175 |  |  | 100 | mA |
| Iout2 | VOUT Current (Battery Back-up) | $\begin{gathered} \mathrm{V}_{\text {OUT2 }}> \\ \mathrm{V}_{\text {BAT }}-0.3 \mathrm{~V} \end{gathered}$ |  |  | 100 |  |  | 100 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Voltage |  | $0.7 \mathrm{~V}_{\mathrm{CC}}$ |  | $V_{C C}+0.3$ | $0.7 \mathrm{~V}_{\mathrm{CC}}$ |  | $\mathrm{V}_{\mathrm{CC}}+0.3$ | V |
| VIL | Input Low Voltage |  | -0.3 |  | $0.3 \mathrm{~V}_{\mathrm{CC}}$ | -0.3 |  | $0.3 \mathrm{~V}_{\mathrm{CC}}$ | V |
| $V_{\text {BAT }}$ | Battery Voltage |  | 2.5 | 3.0 | $3.5{ }^{(9)}$ | 2.5 | 3.0 | $3.5{ }^{(9)}$ | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output High Voltage ${ }^{(6)}$ | $\mathrm{IOH}=-1.0 \mathrm{~mA}$ | 2.4 |  |  | 2.4 |  |  | V |
| $\mathrm{V}_{\mathrm{OHB}}{ }^{(7)}$ | $\mathrm{V}_{\mathrm{OH}}$ (Battery Back-up) | $\begin{aligned} & \text { lout2 }= \\ & -1.0 \mu \mathrm{~A} \end{aligned}$ | 2.5 | 2.9 | 3.5 | 2.5 | 2.9 | 3.5 | V |
| $\mathrm{V}_{\mathrm{OL}}$ | Output Low Voltage | $\mathrm{IOL}=3.0 \mathrm{~mA}$ |  |  | 0.4 |  |  | 0.4 | V |
|  | Output Low Voltage (Open Drain) ${ }^{(8)}$ | $\mathrm{lOL}=10 \mathrm{~mA}$ |  |  | 0.4 |  |  | 0.4 | V |
| VPFD | Power Fail Deselect |  | 4.20 | 4.40 | 4.50 | 2.55 | 2.60 | 2.70 | V |
| VPFI | PFI Input Threshold | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}=5 \mathrm{~V}(\mathrm{Y}) \\ & \mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}(\mathrm{~V}) \end{aligned}$ | 1.225 | 1.250 | 1.275 | 1.225 | 1.250 | 1.275 | V |
|  | PFI Hysteresis | PFI Rising |  | 20 | 70 |  | 20 | 70 | mV |
| Vso | Battery Back-up Switchover |  |  | 2.5 |  |  | 2.5 |  | V |

Note: 1. Valid for Ambient Operating Temperature: $T_{A}=-40$ to $85^{\circ} \mathrm{C}$; $\mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V or 2.7 to 3.6 V (except where noted).
2. Measured with VOUT and $\bar{E}_{\text {CON }}$ open.
3. $\overline{\text { RSTIN1 }}$ and $\overline{\text { RSTIN2 }}$ internally pulled-up to $\mathrm{V}_{\mathrm{CC}}$ through $100 \mathrm{~K} \Omega$ resistor. WDI internally pulled-down to $\mathrm{V}_{\mathrm{SS}}$ through $100 \mathrm{~K} \Omega$ resistor.
4. Outputs Deselected.
5. External SRAM must match RTC SUPERVISOR chip $\mathrm{V}_{\mathrm{CC}}$ specification.
6. For $\overline{\mathrm{PFO}}$ and SQW pins (CMOS).
7. Conditioned output ( $\overline{\mathrm{E}}_{\mathrm{CON}}$ ) can only sustain CMOS leakage current in the battery back-up mode. Higher leakage currents will reduce battery life.
8. For $\overline{\mathrm{RQ}} / \mathrm{FT} / \mathrm{OUT}, \overline{\mathrm{RST}}$ pins (Open Drain): if pulled-up to supply other than $\mathrm{V}_{\mathrm{CC}}$, this supply must be equal to, or less than 3.0 V when $\mathrm{V}_{\mathrm{CC}}=0 \mathrm{~V}$ (during battery back-up mode).
9. For rechargeable back-up, $\mathrm{V}_{\mathrm{BAT}}$ (max) may be considered $\mathrm{V}_{\mathrm{CC}}$.

## OPERATING MODES

The M41ST85Y/W clock operates as a slave device on the serial bus. Access is obtained by implementing a start condition followed by the correct slave address (DOh). The 64 bytes contained in the device can then be accessed sequentially in the following order:

1. Tenths/Hundredths of a Second Register
2. Seconds Register
3. Minutes Register
4. Century/Hours Register
5. Day Register
6. Date Register
7. Month Register
8. Year Register
9. Control Register
10. Watchdog Register

11-16. Alarm Registers
17-19. Reserved
20. Square Wave Register

21-64. User RAM
The M41ST85Y/W clock continually monitors $\mathrm{V}_{\mathrm{CC}}$ for an out-of-tolerance condition. Should $V_{C C}$ fall below VPFD, the device terminates an access in progress and resets the device address counter. Inputs to the device will not be recognized at this time to prevent erroneous data from being written to the device from a an out-of-tolerance system. When $\mathrm{V}_{\mathrm{CC}}$ falls below $\mathrm{V}_{\mathrm{SO}}$, the device automatically switches over to the battery and powers down into an ultra low current mode of operation to
conserve battery life. As system power returns and $\mathrm{V}_{\mathrm{Cc}}$ rises above $\mathrm{V}_{\text {SO }}$, the battery is disconnected, and the power supply is switched to external $\mathrm{V}_{\mathrm{Cc}}$. Write protection continues until $\mathrm{V}_{\mathrm{CC}}$ reaches $V_{\text {PFD }}(\min )$ plus $t_{R E C}(\min )$.
For more information on Battery Storage Life refer to Application Note AN1012.

## 2-Wire Bus Characteristics

The bus is intended for communication between different ICs. It consists of two lines: a bi-directional data signal (SDA) and a clock signal (SCL). Both the SDA and SCL lines must be connected to a positive supply voltage via a pull-up resistor.
The following protocol has been defined:

- Data transfer may be initiated only when the bus is not busy.
- During data transfer, the data line must remain stable whenever the clock line is High.
- Changes in the data line, while the clock line is High, will be interpreted as control signals.
Accordingly, the following bus conditions have been defined:
Bus not busy. Both data and clock lines remain High.
Start data transfer. A change in the state of the data line, from High to Low, while the clock is High, defines the START condition.
Stop data transfer. A change in the state of the data line, from Low to High, while the clock is High, defines the STOP condition.

Data Valid. The state of the data line represents valid data when after a start condition, the data line is stable for the duration of the high period of the clock signal. The data on the line may be changed during the Low period of the clock signal. There is one clock pulse per bit of data.
Each data transfer is initiated with a start condition and terminated with a stop condition. The number of data bytes transferred between the start and stop conditions is not limited. The information is transmitted byte-wide and each receiver acknowledges with a ninth bit.
By definition a device that gives out a message is called "transmitter," the receiving device that gets the message is called "receiver." The device that controls the message is called "master." The devices that are controlled by the master are called "slaves."

Acknowledge. Each byte of eight bits is followed by one Acknowledge Bit. This Acknowledge Bit is a low level put on the bus by the receiver whereas the master generates an extra acknowledge related clock pulse. A slave receiver which is addressed is obliged to generate an acknowledge after the reception of each byte that has been clocked out of the slave transmitter.
The device that acknowledges has to pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is a stable Low during the High period of the acknowledge related clock pulse. Of course, setup and hold times must be taken into account. A master receiver must signal an end of data to the slave transmitter by not generating an acknowledge on the last byte that has been clocked out of the slave. In this case the transmitter must leave the data line High to enable the master to generate the STOP condition.

Figure 9. Serial Bus Data Transfer Sequence


Figure 10. Acknowledgement Sequence


Figure 11. WRITE Cycle Timing: RTC \& External SRAM Control Signals


Figure 12. Bus Timing Requirements Sequence


Table 6. AC Characteristics

| Symbol | Parameter ${ }^{(1)}$ |  | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| fscl | SCL Clock Frequency |  | 0 | 400 | kHz |
| tBuF | Time the bus must be free before a new transmission can start |  | 1.3 |  | $\mu \mathrm{s}$ |
| texpd | $\overline{\mathrm{EX}}$ to $\overline{\mathrm{E}}_{\text {Con }}$ Propagation Delay | M41ST85Y |  | 10 | ns |
|  |  | M41ST85W |  | 15 |  |
| $\mathrm{t}_{\mathrm{F}}$ | SDA and SCL Fall Time |  |  | 300 | ns |
| thd:DAT ${ }^{(2)}$ | Data Hold Time |  | 0 |  | $\mu \mathrm{s}$ |
| thd:STA | START Condition Hold Time (after this period the first clock pulse is generated) |  | 600 |  | ns |
| thigh | Clock High Period |  | 600 |  | ns |
| tLow | Clock Low Period |  | 1.3 |  | $\mu \mathrm{s}$ |
| $t_{R}$ | SDA and SCL Rise Time |  |  | 300 | ns |
| tsu:DAT | Data Setup Time |  | 100 |  | ns |
| tsu:STA | START Condition Setup Time (only relevant for a repeated start condition) |  | 600 |  | ns |
| tsu:sto | STOP Condition Setup Time |  | 600 |  | ns |

Note: 1. Valid for Ambient Operating Temperature: $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V or 2.7 to 3.6 V (except where noted).
2. Transmitter must internally provide a hold time to bridge the undefined region ( 300 ns max) of the falling edge of SCL.

## READ Mode

In this mode the master reads the M41ST85Y/W slave after setting the slave address (see Figure 13, page 13). Following the WRITE Mode Control Bit ( $R / W=0$ ) and the Acknowledge Bit, the word address 'An' is written to the on-chip address pointer. Next the START condition and slave address are repeated followed by the READ Mode Control Bit ( $\mathrm{R} / \overline{\mathrm{W}}=1$ ). At this point the master transmitter becomes the master receiver.
The data byte which was addressed will be transmitted and the master receiver will send an Acknowledge Bit to the slave transmitter. The address pointer is only incremented on reception of an Acknowledge Clock. The M41ST85Y/W slave transmitter will now place the data byte at address $\mathrm{An}+1$ on the bus, the master receiver reads and acknowledges the new byte and the address pointer is incremented to $\mathrm{An}+2$.

This cycle of reading consecutive addresses will continue until the master receiver sends a STOP condition to the slave transmitter (see Figure 14, page 13).
The system-to-user transfer of clock data will be halted whenever the address being read is a clock address ( 00 h to 07 h ). The update will resume either due to a Stop Condition or when the pointer increments to a non-clock or RAM address.
Note: This is true both in READ Mode and WRITE Mode.
An alternate READ Mode may also be implemented whereby the master reads the M41ST85Y/W slave without first writing to the (volatile) address pointer. The first address that is read is the last one stored in the pointer (see Figure 15, page 14).

Figure 13. Slave Address Location


Figure 14. READ Mode Sequence


Figure 15. Alternate READ Mode Sequence


## WRITE Mode

In this mode the master transmitter transmits to the M41ST85Y/W slave receiver. Bus protocol is shown in Figure 16, page 14. Following the START condition and slave address, a logic '0' (R/ $\bar{W}=0$ ) is placed on the bus and indicates to the addressed device that word address An will follow and is to be written to the on-chip address pointer. The data word to be written to the memory is
strobed in next and the internal address pointer is incremented to the next memory location within the RAM on the reception of an acknowledge clock. The M41ST85Y/W slave receiver will send an acknowledge clock to the master transmitter after it has received the slave address (see Figure 13, page 13) and again after it has received the word address and each data byte.

Figure 16. WRITE Mode Sequence


## Data Retention Mode

With valid VCC applied, the M41ST85Y/W can be accessed as described above with READ or WRITE Cycles. Should the supply voltage decay, the M41ST85Y/W will automatically deselect, write protecting itself (and any external SRAM) when $V_{C C}$ falls between $V_{\text {PFD }}(\max )$ and $V_{\text {PFD }}(\min )$. This is accomplished by internally inhibiting access to the clock registers. At this time, the Reset pin ( $\overline{\mathrm{RST}}$ ) is driven active and will remain active until $\mathrm{V}_{\mathrm{CC}}$ returns to nominal levels. External RAM access is inhibited in a similar manner by forcing $\overline{\mathrm{E}}_{\mathrm{CON}}$ to a high level. This level is within 0.2 volts of the $\mathrm{V}_{\text {BAT }}$. E $_{\text {CON }}$ will remain at this level as long as $\mathrm{V}_{\mathrm{CC}}$ remains at an out-of-tolerance condition. When VCC falls below the Battery Back-up Switchover Voltage ( $\mathrm{V}_{\mathrm{SO}}$ ), power input is switched from the $\mathrm{V}_{\mathrm{C}}$ pin to the SNAPHAT ${ }^{\circledR}$ battery, and the clock registers and external SRAM are maintained from the attached battery supply.
All outputs become high impedance. The $\mathrm{V}_{\text {Out }}$ pin is capable of supplying $100 \mu \mathrm{~A}$ of current to the attached memory with less than 0.3 volts drop under this condition. On power up, when $\mathrm{V}_{\mathrm{CC}}$ returns to a nominal value, write protection continues for $t_{\text {REC }}$ by inhibiting $\bar{E}_{\text {CON }}$. The RST signal also remains active during this time (see Figure 17, page 16).

Note: Most low power SRAMs on the market today can be used with the M41ST85Y/W RTC SUPERVISOR. There are, however some criteria which should be used in making the final choice of
an SRAM to use. The SRAM must be designed in a way where the chip enable input disables all other inputs to the SRAM. This allows inputs to the M41ST85Y/W and SRAMs to be "Don't Care" once $\mathrm{V}_{\mathrm{CC}}$ falls below $\mathrm{V}_{\text {PFD }}(\mathrm{min})$. The SRAM should also guarantee data retention down to $\mathrm{V}_{\mathrm{CC}}=2.0$ volts. The chip enable access time must be sufficient to meet the system needs with the chip enable output propagation delays included. If the SRAM includes a second chip enable pin (E2), this pin should be tied to Vout.
If data retention lifetime is a critical parameter for the system, it is important to review the data retention current specifications for the particular SRAMs being evaluated. Most SRAMs specify a data retention current at 3.0 volts. Manufacturers generally specify a typical condition for room temperature along with a worst case condition (generally at elevated temperatures). The system level requirements will determine the choice of which value to use. The data retention current value of the SRAMs can then be added to the I ${ }_{\text {BAT }}$ value of the M41ST85Y/W to determine the total current requirements for data retention. The available battery capacity for the SNAPHAT ${ }^{\circledR}$ of your choice can then be divided by this current to determine the amount of data retention available (see Table 15 , page 27 ).
For a further more detailed review of lifetime calculations, please see Application Note AN1012.

Figure 17. Power Down/Up Mode AC Waveforms


Table 7. Power Down/Up AC Characteristics

| Symbol | Parameter ${ }^{(1)}$ | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{tF}^{(2)}$ | $\mathrm{V}_{\text {PFD }}(\mathrm{max})$ to $\mathrm{V}_{\text {PFD }}(\min ) \mathrm{V}_{\text {CC }}$ Fall Time | 300 |  |  | $\mu \mathrm{S}$ |
| $\mathrm{t}_{\mathrm{FB}}{ }^{(3)}$ | $\mathrm{V}_{\text {PFD }}(\mathrm{min})$ to $\mathrm{V}_{\text {SS }} \mathrm{V}_{\text {CC }}$ Fall Time | 10 |  |  | $\mu \mathrm{s}$ |
| tpd | $\overline{E X}$ at $\mathrm{V}_{\mathbb{H}}$ before Power Down | 0 |  |  | $\mu \mathrm{s}$ |
| tpfo | PFI to $\overline{\text { PFO }}$ Propagation Delay |  | 15 | 25 | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{R}}$ | $\mathrm{V}_{\text {PFD }}(\mathrm{min})$ to $\mathrm{V}_{\text {PFD }}(\mathrm{max}) \mathrm{V}_{\text {cC }}$ Rise Time | 10 |  |  | $\mu \mathrm{s}$ |
| trB | $\mathrm{V}_{\text {SS }}$ to V $\mathrm{VFFD}^{\text {(min) }}$ V $\mathrm{V}_{\text {c }}$ Rise Time | 1 |  |  | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {REC }}{ }^{(4)}$ | Power up Deselect Time | 40 |  | 200 | ms |

Note: 1. Valid for Ambient Operating Temperature: $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C}$; $\mathrm{V}_{\mathrm{CC}}=4.5$ to 5.5 V or 2.7 to 3.6 V (except where noted).
2. $\mathrm{V}_{\text {PFD }}(\max )$ to $\mathrm{V}_{\text {PFD }}(\min )$ fall time of less than $t_{F}$ may result in deselection/write protection not occurring until $200 \mu \mathrm{~s}$ after $\mathrm{V}_{\text {CC }}$ passes $\mathrm{V}_{\text {PFD }}(\mathrm{min})$.
3. $\mathrm{V}_{\text {PFD }}(\mathrm{min})$ to $\mathrm{V}_{S S}$ fall time of less than tFB may cause corruption of RAM data.
4. Programmable (see Table 12, page 25)

## CLOCK OPERATION

The eight byte clock register (see Table 8, page 18) is used to both set the clock and to read the date and time from the clock, in a binary coded decimal format. Tenths/Hundredths of Seconds, Seconds, Minutes, and Hours are contained within the first four registers.
Note: A WRITE to any clock register will result in the Tenths/Hundredths of Seconds being reset to " 00 ," and Tenths/Hundredths of Seconds cannot be written to any value other than " 00 ."
Bits D6 and D7 of Clock Register 03h (Century/ Hours Register) contain the CENTURY ENABLE Bit (CEB) and the CENTURY Bit (CB). Setting CEB to a ' 1 ' will cause CB to toggle, either from ' 0 ' to ' 1 ' or from ' 1 ' to ' 0 ' at the turn of the century (depending upon its initial state). If CEB is set to a ' 0 ,' CB will not toggle. Bits D0 through D2 of Register 04h contain the Day (day of week). Registers 05h, 06h, and 07h contain the Date (day of month), Month and Years. The ninth clock register is the Control Register (this is described in the Clock Calibration section). Bit D7 of Register 01h contains the STOP Bit (ST). Setting this bit to a ' 1 ' will cause the oscillator to stop. If the device is expected to spend a significant amount of time on the shelf, the oscillator may be stopped to reduce current drain. When reset to a ' 0 ' the oscillator restarts within one second.
The eight Clock Registers may be read one byte at a time, or in a sequential block. The Control Register (Address location 08h) may be accessed independently. Provision has been made to assure that a clock update does not occur while any of the
eight clock addresses are being read. If a clock address is being read, an update of the clock registers will be halted. This will prevent a transition of data during the READ.
Note: When a power failure occurs, the Halt Update Bit (HT) will automatically be set to a '1.' This will prevent the clock from updating the TIMEKEEPER ${ }^{\circledR}$ registers, and will allow the user to read the exact time of the power-down event. Resetting the HT Bit to a '0' will allow the clock to update the TIMEKEEPER registers with the current time.

## TIMEKEEPER ${ }^{\circledR}$ Registers

The M41ST85Y/W offers 20 internal registers which contain Clock, Alarm, Watchdog, Flag, Square Wave and Control data. These registers are memory locations which contain external (user accessible) and internal copies of the data (usually referred to as BiPORT ${ }^{\text {MT }}$ TIMEKEEPER cells). The external copies are independent of internal functions except that they are updated periodically by the simultaneous transfer of the incremented internal copy. The internal divider (or clock) chain will be reset upon the completion of a WRITE to any clock address.
The system-to-user transfer of clock data will be halted whenever the address being read is a clock address ( 00 h to 07 h ). The update will resume either due to a Stop Condition or when the pointer increments to a non-clock or RAM address.
TIMEKEEPER and Alarm Registers store data in BCD. Control, Watchdog and Square Wave Registers store data in Binary Format.

Table 8. TIMEKEEPER ${ }^{\circledR}$ Register Map

| Address | Data |  |  |  |  |  |  |  | Function/Range BCD Format |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |
| 00h | 0.1 Seconds |  |  |  | 0.01 Seconds |  |  |  | Seconds | 00-99 |
| 01h | ST | 10 Seconds |  |  | Seconds |  |  |  | Seconds | 00-59 |
| 02h | 0 | 10 Minutes |  |  | Minutes |  |  |  | Minutes | 00-59 |
| 03h | CEB | CB | 10 Hours |  | Hours (24 Hour Format) |  |  |  | Century/Hours | 0-1/00-23 |
| 04h | TR | 0 | 0 | 0 | 0 | Day of Week |  |  | Day | 01-7 |
| 05h | 0 | 0 | 10 Date |  | Date: Day of Month |  |  |  | Date | 01-31 |
| 06h | 0 | 0 | 0 | 10M | Month |  |  |  | Month | 01-12 |
| 07h | 10 Years |  |  |  | Year |  |  |  | Year | 00-99 |
| 08h | OUT | FT | S | Calibration |  |  |  |  | Control |  |
| 09h | WDS | BMB4 | BMB3 | BMB2 | BMB1 | BMB0 | RB1 | RB0 | Watchdog |  |
| 0Ah | AFE | SQWE | ABE | Al 10M | Alarm Month |  |  |  | Al Month | 01-12 |
| 0Bh | RPT4 | RPT5 | Al 10 Date |  | Alarm Date |  |  |  | Al Date | 01-31 |
| 0Ch | RPT3 | HT | AI 10 Hour |  | Alarm Hour |  |  |  | Al Hour | 00-23 |
| 0Dh | RPT2 | Alarm 10 Minutes |  |  | Alarm Minutes |  |  |  | Al Min | 00-59 |
| 0Eh | RPT1 | Alarm 10 Seconds |  |  | Alarm Seconds |  |  |  | Al Sec | 00-59 |
| 0Fh | WDF | AF | 0 | BL | 0 | 0 | 0 | 0 | Flags |  |
| 10h | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Reserved |  |
| 11h | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Reserved |  |
| 12h | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Reserved |  |
| 13h | RS3 | RS2 | RS1 | RS0 | 0 | 0 | 0 | 0 | SQW |  |

Keys: S = Sign Bit
FT = Frequency Test Bit
ST = Stop Bit
0 = Must be set to zero
BL = Battery Low Flag (Read only)
BMB0-BMB4 = Watchdog Multiplier Bits
CEB = Century Enable Bit
$C B=$ Century Bit
OUT = Output level
AFE = Alarm Flag Enable Flag

RB0-RB1 = Watchdog Resolution Bits
WDS = Watchdog Steering Bit
ABE = Alarm in Battery Back-Up Mode Enable Bit
RPT1-RPT5 = Alarm Repeat Mode Bits
WDF = Watchdog flag (Read only)
AF = Alarm flag (Read only)
SQWE = Square Wave Enable
RS0-RS3 = SQW Frequency
HT = Halt Update Bit
$T R=t_{\text {REC }}$

## Calibrating the Clock

The M41ST85Y/W is driven by a quartz controlled oscillator with a nominal frequency of $32,768 \mathrm{~Hz}$. The devices are tested not exceed +/-35 PPM (parts per million) oscillator frequency error at $25^{\circ} \mathrm{C}$, which equates to about $+/-1.53$ minutes per month. When the Calibration circuit is properly employed, accuracy improves to better than $+1 /-2$ ppm at $25^{\circ} \mathrm{C}$.
The oscillation rate of crystals changes with temperature (see Figure 21, page 26). Therefore, the M41ST85Y/W design employs periodic counter correction. The calibration circuit adds or subtracts counts from the oscillator divider circuit at the divide by 256 stage, as shown in Figure 22, page 26. The number of times pulses which are blanked (subtracted, negative calibration) or split (added, positive calibration) depends upon the value loaded into the five Calibration Bits found in the Control Register. Adding counts speeds the clock up, subtracting counts slows the clock down.
The Calibration Bits occupy the five lower order bits (D4-D0) in the Control Register (08h). These bits can be set to represent any value between 0 and 31 in binary form. Bit D5 is a Sign Bit; ' 1 ' indicates positive calibration, '0' indicates negative calibration. Calibration occurs within a 64 minute cycle. The first 62 minutes in the cycle may, once per minute, have one second either shortened by 128 or lengthened by 256 oscillator cycles. If a binary ' 1 ' is loaded into the register, only the first 2 minutes in the 64 minute cycle will be modified; if a binary 6 is loaded, the first 12 will be affected, and so on.
Therefore, each calibration step has the effect of adding 512 or subtracting 256 oscillator cycles for every $125,829,120$ actual oscillator cycles, that is +4.068 or -2.034 PPM of adjustment per calibration step in the calibration register. Assuming that the oscillator is running at exactly $32,768 \mathrm{~Hz}$, each of the 31 increments in the Calibration byte would represent +10.7 or -5.35 seconds per month which corresponds to a total range of +5.5 or -2.75 minutes per month.
Two methods are available for ascertaining how much calibration a given M41ST85Y/W may require.
The first involves setting the clock, letting it run for a month and comparing it to a known accurate reference and recording deviation over a fixed period of time. Calibration values, including the number of seconds lost or gained in a given period, can be found in Application Note AN934, "TIMEKEEP$E R^{\circledR}$ CALIBRATION." This allows the designer to give the end user the ability to calibrate the clock as the environment requires, even if the final prod-
uct is packaged in a non-user serviceable enclosure. The designer could provide a simple utility that accesses the Calibration byte.
The second approach is better suited to a manufacturing environment, and involves the use of the IRQ/FT/OUT pin. The pin will toggle at 512 Hz , when the Stop Bit (ST, D7 of 01h) is ' 0 ,' the Frequency Test Bit (FT, D6 of 08h) is '1,' the Alarm Flag Enable Bit (AFE, D7 of OAh) is '0,' and the Watchdog Steering Bit (WDS, D7 of 09h) is '1' or the Watchdog Register $(09 \mathrm{~h}=0)$ is reset.
Any deviation from 512 Hz indicates the degree and direction of oscillator frequency shift at the test temperature. For example, a reading of 512.010124 Hz would indicate a +20 PPM oscillator frequency error, requiring a -10 (XX001010) to be loaded into the Calibration Byte for correction. Note that setting or changing the Calibration Byte does not affect the Frequency test output frequency.
The $\overline{\mathrm{IRQ}} / \mathrm{FT} / \mathrm{OUT}$ pin is an open drain output which requires a pull-up resistor to $\mathrm{V}_{\mathrm{CC}}$ for proper operation. A 500 to 10 k resistor is recommended in order to control the rise time. The FT Bit is cleared on power-down.

## Setting Alarm Clock Registers

Address locations 0Ah-0Eh contain the alarm settings. The alarm can be configured to go off at a prescribed time on a specific month, date, hour, minute, or second, or repeat every year, month, day, hour, minute, or second. It can also be programmed to go off while the M41ST85Y/W is in the battery back-up to serve as a system wake-up call. Bits RPT5-RPT1 put the alarm in the repeat mode of operation. Table 9, page 20 shows the possible configurations. Codes not listed in the table default to the once per second mode to quickly alert the user of an incorrect alarm setting.
When the clock information matches the alarm clock settings based on the match criteria defined by RPT5-RPT1, the AF (Alarm Flag) is set. If AFE (Alarm Flag Enable) is also set, the alarm condition activates the $\overline{\mathrm{RQ}} / \mathrm{FT} / \mathrm{OUT}$ pin as shown in Figure 18, page 20. To disable alarm, write '0' to the Alarm Date Register and to RPT5-RPT1.
Note: If the address pointer is allowed to increment to the Flag Register address, an alarm condition will not cause the Interrupt/Flag to occur until the address pointer is moved to a different address. It should also be noted that if the last address written is the "Alarm Seconds," the address pointer will increment to the Flag address, causing this situation to occur.

The $\overline{\mathrm{RQ}} / \mathrm{FT} / \mathrm{OUT}$ output is cleared by a READ to the Flags Register. A subsequent READ of the Flags Register is necessary to see that the value of the Alarm Flag has been reset to ' 0 .'
The $\overline{\mathrm{RQ}} / \mathrm{FT} / \mathrm{OUT}$ pin can also be activated in the battery back-up mode. The IRQ/FT/OUT will go low if an alarm occurs and both ABE (Alarm in Battery Back-up Mode Enable) and AFE are set. The
$A B E$ and AFE Bits are reset during power-up, therefore an alarm generated during power-up will only set AF. The user can read the Flag Register at system boot-up to determine if an alarm was generated while the M41ST85Y/W was in the deselect mode during power-up. Figure 19, page 21 illustrates the back-up mode alarm timing.

Figure 18. Alarm Interrupt Reset Waveform


Table 9. Alarm Repeat Modes

| RPT5 | RPT4 | RPT3 | RPT2 | RPT1 | Alarm Setting |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 1 | 1 | 1 | 1 | Once per Second |
| 1 | 1 | 1 | 1 | 0 | Once per Minute |
| 1 | 1 | 1 | 0 | 0 | Once per Hour |
| 1 | 1 | 0 | 0 | 0 | Once per Day |
| 1 | 0 | 0 | 0 | 0 | Once per Month |
| 0 | 0 | 0 | 0 | 0 | Once per Year |

Figure 19. Back-Up Mode Alarm Waveform


## Watchdog Timer

The watchdog timer can be used to detect an out-of-control microprocessor. The user programs the watchdog timer by setting the desired amount of time-out into the Watchdog Register, address 09h. Bits BMB4-BMB0 store a binary multiplier and the two lower order bits RB1-RB0 select the resolution, where $00=1 / 16$ second, $01=1 / 4$ second, $10=1$ second, and $11=4$ seconds. The amount of timeout is then determined to be the multiplication of the five-bit multiplier value with the resolution. (For example: writing 00001110 in the Watchdog Register $=3^{* 1}$ or 3 seconds).
Note: The accuracy of the timer is within $\pm$ the selected resolution.
If the processor does not reset the timer within the specified period, the M41ST85Y/W sets the WDF (Watchdog Flag) and generates a watchdog interrupt or a microprocessor reset.
The most significant bit of the Watchdog Register is the Watchdog Steering Bit (WDS). When set to a ' 0 ,' the watchdog will activate the IRQ/FT/OUT pin when timed-out. When WDS is set to a ' 1, ' the watchdog will output a negative pulse on the RST pin for trec. The Watchdog register, FT, AFE, ABE and SQWE Bits will reset to a ' 0 ' at the end of a Watchdog time-out when the WDS Bit is set to a '1.'

The watchdog timer can be reset by two methods: 1) a transition (high-to-low or low-to-high) can be applied to the Watchdog Input pin (WDI) or 2) the microprocessor can perform a WRITE of the Watchdog Register. The time-out period then starts over.
Note: The WDI pin should be tied to $\mathrm{V}_{\mathrm{SS}}$ if not used.
In order to perform a software reset of the watchdog timer, the original time-out period can be written into the Watchdog Register, effectively restarting the count-down cycle.
Should the watchdog timer time-out, and the WDS Bit is programmed to output an interrupt, a value of 00h needs to be written to the Watchdog Register in order to clear the IRQ/FT/OUT pin. This will also disable the watchdog function until it is again programmed correctly. A READ of the Flags Register will reset the Watchdog Flag (Bit D7; Register $0 F h$ ).
The watchdog function is automatically disabled upon power-up and the Watchdog Register is cleared. If the watchdog function is set to output to the $\overline{\mathrm{RQ}} / \mathrm{FT} / \mathrm{OUT}$ pin and the frequency test function is activated, the watchdog function prevails and the frequency test function is denied.

## Square Wave Output

The M41ST85Y/W offers the user a programmable square wave function which is output on the SQW pin. RS3-RS0 bits located in 13h establish the square wave output frequency. These frequencies are listed in Table 10. Once the selection
of the SQW frequency has been completed, the SQW pin can be turned on and off under software control with the Square Wave Enable Bit (SQWE) located in Register OAh.

Table 10. Square Wave Output Frequency

| Square Wave Bits |  |  |  | Square Wave |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RS3 | RS2 | RS1 | RS0 | Frequency | Units |
| 0 | 0 | 0 | 0 | None | - |
| 0 | 0 | 0 | 1 | 32.768 | kHz |
| 0 | 0 | 1 | 0 | 8.192 | kHz |
| 0 | 0 | 1 | 1 | 4.096 | kHz |
| 0 | 1 | 0 | 0 | 2.048 | kHz |
| 0 | 1 | 0 | 1 | 1.024 | kHz |
| 0 | 1 | 1 | 0 | 512 | Hz |
| 0 | 1 | 1 | 1 | 256 | Hz |
| 1 | 0 | 0 | 0 | 128 | Hz |
| 1 | 0 | 0 | 1 | 64 | Hz |
| 1 | 0 | 1 | 0 | 32 | Hz |
| 1 | 0 | 1 | 1 | 16 | Hz |
| 1 | 1 | 0 | 0 | 8 | Hz |
| 1 | 1 | 0 | 1 | 4 | Hz |
| 1 | 1 | 1 | 0 | 2 | Hz |
| 1 | 1 | 1 | 1 | 1 | Hz |

## Power-on Reset

The M41ST85Y/W continuously monitors $V_{C c}$. When Vcc falls to the power fail detect trip point, the RST pulls low (open drain) and remains low on power-up for $t_{\text {REC }}$ after $\mathrm{V}_{\text {CC }}$ passes $\mathrm{V}_{\text {PFD }}(m a x)$. The RST pin is an open drain output and an appropriate pull-up resistor should be chosen to control rise time.

## Reset Inputs ( $\overline{\text { RSTIN1 }} \& \overline{\text { RSTIN2 }}$ )

The M41ST85Y/W provides two independent inputs which can generate an output reset. The duration and function of these resets is identical to a reset generated by a power cycle. Table 11 and Figure 20 illustrate the AC reset characteristics of this function. Pulses shorter than trLRH1 and $t_{\text {RLRH } 2 \text { will not }}$ generate a reset condition. RSTIN1 and RSTIN2 are each internally pulled up to $\mathrm{V}_{\mathrm{CC}}$ through a $100 \mathrm{k} \Omega$ resistor.

Figure 20. $\overline{\text { RSTIN1 }} \& \overline{\text { RSTIN2 }}$ Timing Waveforms


Note: With pull-up resistor
Table 11. Reset AC Characteristics

| Symbol | Parameter $^{(1)}$ | Min | Max | Unit |
| :---: | :--- | :---: | :---: | :---: |
| trLRH1 $^{(2)}$ | $\overline{\text { RSTIN1 }}$ Low to $\overline{\text { RSTIN1 }}$ High | 200 |  | ns |
| trLRH2 $^{(3)}$ | $\overline{\text { RSTIN2 }}$ Low to $\overline{\text { RSTIN2 }}$ High | 100 |  | ms |
| tr1HRH $^{(4)}$ | $\overline{\text { RSTIN1 }}$ High to $\overline{\text { RST }}$ High | 40 | 200 | ms |
| tr2HRH $^{(4)}$ | $\overline{\text { RSTIN2 }}$ High to $\overline{\text { RST }}$ High | 40 | 200 | ms |

Note: 1. Valid for Ambient Operating Temperature: $T_{A}=-40$ to $85^{\circ} \mathrm{C}$; $\mathrm{V}_{C C}=4.5$ to 5.5 V or 2.7 to 3.6 V (except where noted).
2. Pulse width less than 50 ns will result in no RESET (for noise immunity).
3. Pulse width less than 20 ms will result in no RESET (for noise immunity).
4. Programmable (see Table 12, page 25).

## Power-fail INPUT/OUTPUT

The Power-Fail Input (PFI) is compared to an internal reference voltage ( 1.25 V ). If PFI is less than the power-fail threshold (VPFI), the Power-Fail Output (PFO) will go low. This function is intended for use as an undervoltage detector to signal a failing power supply. Typically PFI is connected through an external voltage divider (see Figure 7, page 7) to either the unregulated DC input (if it is available) or the regulated output of the $\mathrm{V}_{\mathrm{CC}}$ regulator. The voltage divider can be set up such that the voltage at PFI falls below $\mathrm{V}_{\mathrm{PFI}}$ several milliseconds before the regulated $V_{C C}$ input to the M41ST85Y/W or the microprocessor drops below the minimum operating voltage.
During battery back-up, the power-fail comparator turns off and PFO goes (or remains) low. This occurs after $\mathrm{V}_{\mathrm{CC}}$ drops below $\mathrm{V}_{\text {PFD }}(\mathrm{min})$. When power returns, PFO is forced high, irrespective of $\mathrm{V}_{\mathrm{PFI}}$ for the write protect time (tREC), which is the time from $V_{\text {PFD }}$ (max) until the inputs are recognized. At the end of this time, the power-fail comparator is enabled and PFO follows PFI. If the comparator is unused, PFI should be connected to $\mathrm{V}_{\text {SS }}$ and PFO left unconnected.

## Century Bit

Bits D7 and D6 of Clock Register 03h contain the CENTURY ENABLE Bit (CEB) and the CENTURY Bit (CB). Setting CEB to a '1' will cause CB to toggle, either from a ' 0 ' to ' 1 ' or from ' 1 ' to ' 0 ' at the turn of the century (depending upon its initial state). If CEB is set to a ' 0 ,' CB will not toggle.

## Output Driver Pin

When the FT Bit, AFE Bit and watchdog register are not set, the $\overline{\mathrm{IRQ}} / \mathrm{FT} /$ OUT pin becomes an output driver that reflects the contents of D7 of the Control Register. In other words, when D7 (OUT Bit) and D6 (FT Bit) of address location 08h are a ' 0 ,' then the $\overline{\mathrm{IRQ}} / \mathrm{FT} / \mathrm{OUT}$ pin will be driven low.
Note: The $\overline{\mathrm{RQ}} / \mathrm{FT} / \mathrm{OUT}$ pin is an open drain which requires an external pull-up resistor.

## Battery Low Warning

The M41ST85Y/W automatically performs battery voltage monitoring upon power-up and at factoryprogrammed time intervals of approximately 24 hours. The Battery Low (BL) Bit, Bit D4 of Flags Register 0Fh, will be asserted if the battery voltage is found to be less than approximately 2.5 V . The

BL Bit will remain asserted until completion of battery replacement and subsequent battery low monitoring tests, either during the next power-up sequence or the next scheduled 24-hour interval.
If a battery low is generated during a power-up sequence, this indicates that the battery is below approximately 2.5 volts and may not be able to maintain data integrity in the SRAM. Data should be considered suspect and verified as correct. A fresh battery should be installed.
If a battery low indication is generated during the 24-hour interval check, this indicates that the battery is near end of life. However, data is not compromised due to the fact that a nominal $\mathrm{V}_{\mathrm{Cc}}$ is supplied. In order to insure data integrity during subsequent periods of battery back-up mode, the battery should be replaced. The SNAPHAT top may be replaced while $\mathrm{V}_{\mathrm{CC}}$ is applied to the device.
Note: This will cause the clock to lose time during the interval the SNAPHAT battery/crystal top is disconnected.
The M41ST85Y/W only monitors the battery when a nominal $\mathrm{V}_{\mathrm{Cc}}$ is applied to the device. Thus applications which require extensive durations in the battery back-up mode should be powered-up periodically (at least once every few months) in order for this technique to be beneficial. Additionally, if a battery low is indicated, data integrity should be verified upon power-up via a checksum or other technique.

## $t_{\text {REC }}$ Bit

Bit D7 of Clock Register 04h contains the trec Bit (TR). thec refers to the automatic continuation of the deselect time after $\mathrm{V}_{\text {CC }}$ reaches $\mathrm{V}_{\text {PFD }}$. This allows for a voltage settling time before WRITEs may again be performed to the device after a pow-er-down condition. The trec Bit will allow the user to set the length of this deselect time as defined by Table 12, page 25.

## Initial Power-on Defaults

Upon initial application of power to the device, the following register bits are set to a '0' state: Watchdog Register, FT, AFE, ABE, SQWE, and TR. The following bits are set to a '1' state: ST, OUT, and HT (see Table 13, page 25).

Table 12. t $_{\text {REC }}$ Definitions

| trec Bit (TR) | STOP Bit (ST) | trec Time $^{*}$ |  | Units |
| :---: | :---: | :---: | :---: | :---: |
|  |  | Min | 98 |  |
| 0 | 0 | 96 | $200^{(1)}$ | ms |
| 0 | 1 | 40 | 2000 | $\mu \mathrm{~s}$ |
| 1 | X | 50 |  |  |

Note: 1. Default Setting
Table 13. Default Values

| Condition | TR | ST | HT | Out | FT | AFE | ABE | SQWE | WATCHDOG <br> Register |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| (1) |  |  |  |  |  |  |  |  |  |$|$

Note: 1. WDS, BMB0-BMB4, RB0, RB1.
2. State of other control bits undefined.
3. $U C=$ Unchanged

Figure 21. Crystal Accuracy Across Temperature


Figure 22. Calibration Waveform


## PART NUMBERING

Table 14. Ordering Information Scheme

$6=-40$ to $85^{\circ} \mathrm{C}$

## Shipping Method for SOIC

blank = Tubes
TR = Tape \& Reel

Note: 1. The 28-pin SOIC package (SOH28) requires the battery/crystal package (SNAPHAT ${ }^{\circledR}$ ) which is ordered separately under the part number "M4TXX-BR12SHX" in plastic tube or "M4TXX-BR12SHXTR" in Tape \& Reel form.
2. The SOX28 package includes an embedded $32,768 \mathrm{~Hz}$ crystal.

Caution: Do not place the SNAPHAT battery package "M4TXX-BR12SH" in conductive foam as it will drain the lithium button-cell battery.
For a list of available options (e.g., Speed, Package) or for further information on any aspect of this device, please contact the ST Sales Office nearest to you.

Table 15. SNAPHAT Battery Table

| Part Number | Description | Package |
| :---: | :--- | :---: |
| M4T28-BR12SH | Lithium Battery $(48 \mathrm{mAh})$ and Crystal SNAPHAT | SH |
| M4T32-BR12SH | Lithium Battery $(120 \mathrm{mAh})$ and Crystal SNAPHAT | SH |

## PACKAGE MECHANICAL INFORMATION

Figure 23. SOH28-28-lead Plastic Small Outline, Battery SNAPHAT, Package Outline


SOH-A

Note: Drawing is not to scale.
Table 16. SOH28-28-lead Plastic Small Outline, battery SNAPHAT, Package Mechanical Data

| Symbol | millimeters |  |  | inches |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Typ | Min | Max | Typ | Min | Max |
| A |  |  | 3.05 |  |  | 0.120 |
| A1 |  | 0.05 | 0.36 |  | 0.002 | 0.014 |
| A2 |  | 2.34 | 2.69 |  | 0.092 | 0.106 |
| B |  | 0.36 | 0.51 |  | 0.014 | 0.020 |
| C |  | 0.15 | 0.32 |  | 0.006 | 0.012 |
| D |  | 17.71 | 18.49 |  | 0.697 | 0.728 |
| E |  | 8.23 | 8.89 |  | 0.324 | 0.350 |
| e | 1.27 | - | - | 0.050 | - | - |
| eB |  | 3.20 | 3.61 |  | 0.126 | 0.142 |
| H |  | 11.51 | 12.70 |  | 0.453 | 0.500 |
| L |  | 0.41 | 1.27 |  | 0.016 | 0.050 |
| $\alpha$ |  | $0^{\circ}$ | $8^{\circ}$ |  | $0^{\circ}$ | $8^{\circ}$ |
| N | 28 |  |  | 28 |  |  |
| CP |  |  | 0.10 |  |  | 0.004 |

Figure 24. SH - 4-pin SNAPHAT Housing for 48mAh Battery \& Crystal, Package Outline


Note: Drawing is not to scale.
Table 17. SH - 4-pin SNAPHAT Housing for 48mAh Battery \& Crystal, Package Mechanical Data

| Symbol | millimeters |  |  | inches |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Typ | Min | Max | Typ | Min | Max |
| A |  |  | 9.78 |  |  | 0.3850 |
| A1 |  | 6.73 | 7.24 |  | 0.2551 | 0.2752 |
| A2 |  | 6.48 | 6.99 |  | 0.0181 | 0.0220 |
| A3 |  | 0.46 | 0.56 |  | 0.8350 | 0.8598 |
| B |  | 21.21 | 21.84 |  | 0.5598 | 0.5902 |
| D |  | 14.22 | 14.99 |  | 0.6122 | 0.6280 |
| E |  | 15.55 | 15.95 |  | 0.1260 | 0.1421 |
| eA |  | 3.20 | 3.61 |  | 0.0799 | 0.0902 |
| eB |  | 2.03 | 2.29 |  |  |  |

Figure 25. SH - 4-pin SNAPHAT Housing for 120mAh Battery \& Crystal, Package Outline


Note: Drawing is not to scale.
Table 18. SH - 4-pin SNAPHAT Housing for 120mAh Battery \& Crystal, Package Mechanical Data

| Symbol | millimeters |  |  | inches |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Typ | Min | Max | Typ | Min | Max |
| A |  |  | 10.54 |  |  | 0.4150 |
| A1 |  | 6.73 | 7.24 |  | 0.2551 | 0.2752 |
| A2 |  | 6.48 | 6.99 |  | 0.0181 | 0.0220 |
| A3 |  | 0.46 | 0.56 |  | 0.8350 | 0.8598 |
| B |  | 21.21 | 21.84 |  | 0.5598 | 0.5902 |
| D |  | 14.22 | 14.99 |  | 0.6122 | 0.6280 |
| E |  | 15.55 | 15.95 |  | 0.1260 | 0.1421 |
| eA |  | 3.20 | 3.61 |  | 0.0799 | 0.0902 |
| eB |  | 2.03 | 2.29 |  |  |  |

Figure 26. SOX28-28-lead Plastic Small Outline, 300mils, Embedded Crystal, Package Outline


Note: Drawing is not to scale.
Table 19. SOX28-28-lead Plastic Small Outline, 300mils, Embedded Crystal, Package Mechanical

| Symbol | millimeters |  |  | inches |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Typ | Min | Max | Typ | Min | Max |
| A |  | 2.44 | 2.69 |  | 0.096 | 0.106 |
| A1 |  | 0.15 | 0.31 |  | 0.006 | 0.012 |
| A2 |  | 2.29 | 2.39 |  | 0.090 | 0.094 |
| B | 0.41 | 0.51 |  | 0.016 | 0.020 |  |
| C | 0.20 | 0.31 |  | 0.008 | 0.012 |  |
| D |  | 17.91 | 18.01 |  | 0.705 | 0.709 |
| ddd |  | 7.57 | 7.67 |  | 0.298 | 0.302 |
| E |  | - | - | 0.050 | - | 0.400 |
| P | 1.27 | 10.16 | 10.52 |  | 0.020 | 0.032 |
| L |  | 0.51 | 0.81 |  | $0^{\circ}$ | $8^{\circ}$ |
| $\alpha$ | $0^{\circ}$ | $8^{\circ}$ |  | 28 |  |  |
| N |  | 28 |  |  |  |  |

## REVISION HISTORY

Table 20. Document Revision History

| Date | Rev. \# | Revision Details |
| :---: | :---: | :--- |
| August 2000 | 1.0 | First issue |
| 24-Aug-00 | 1.1 | Block Diagram added (Figure 3) |
| 12-Oct-00 | 1.2 | trec Table removed, cross references corrected |
| 18-Dec-00 | 2.0 | Reformatted, TOC added, and PFI Input Leakage Current added (Table 5) |
| 18-Jun-01 | 2.1 | Addition of trec information, table changed, one added (Tables 8, 12); changed PFI/PFO <br> graphic (see Figure 6); change to DC and AC Characteristics, Order Information (Tables 5, <br> 6, 14); note added to "Setting Alarm Clock Registers" section; added temp./voltage info. to <br> tables (Table 4, 5, 6, 6, 7); addition of Default Values (Table 13) |
| 22-Jun-01 | 2.2 | Note added to Clock Operation section |
| 26-Jul-01 | 3.0 | Change in Product Maturity |
| 07-Aug-01 | 3.1 | Improve text in "Setting the Alarm Clock" section |
| 20-Aug-01 | 3.2 | Change VPFD values in document |
| 06-Sep-01 | 3.3 | DC Characteristics VBAT changed; VOHB changed; PFI Hysteresis (PFI Rising) spec. <br> added; and Crystal Electrical Characteristics table removed (Tables 5, 6) |
| 03-Dec-01 | 3.4 | Changed READ/WRITE Mode Sequences (Figure 14, 16); change in VPFD lower limit for <br> 5V (M41ST85Y) part only (Table 5, 14) |
| 01-May-02 | 3.5 | Change trec Definition (Table 12); modify reflow time and temperature footnote (Table 2) |
| 03-Jul-02 | 3.6 | Modify DC Characteristics table footnote, Default Values (Tables 5, 13) |
| 15-Nov-02 | 3.7 | Added embedded crystal (MX) package option; corrected initial power-up condition (Figure <br> 2, 3, 5, 6, 7, 26, Table 1, 13, 14, 19) |
| 24-Jan-03 | 3.8 | Update diagrams (Figure 6, 7, 26); update values (Table 7, 11, 12, 13, 19) |
| 25-Feb-03 | 4.0 | New Si changes (Table 7, 11, 12); corrected dimensions (Figure 26; Table 19) |

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